### 2.5 KV Isolated RS-485 Transceivers with Integrated Transformer Driver

## ADM2482E/ADM2487E

## FEATURES

Isolated RS-485/RS-422 transceivers, configurable as half duplex or full duplex
Integrated oscillator driver for external transformer $\pm 15$ kV ESD protection on RS-485 input/output pins
Complies with TIA/EIA-485-A-98 and ISO 8482:1987(E)
Data rate: 500 kbps/16 Mbps
5 V or 3.3 V operation ( $\mathrm{V}_{\mathrm{DD} 1}$ )
256 nodes on bus
True fail-safe receiver inputs
2500 V rms isolation for 1 minute
Reinforced insulation 560 V peak
High common-mode transient immunity: > $\mathbf{2 5} \mathbf{~ k V / \mu s}$
Thermal shutdown protection
Operating temperature range: $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Wide-body, 16-lead SOIC package

## APPLICATIONS

Isolated RS-485/RS-422 interfaces
Industrial field networks
Multipoint data transmission systems

## GENERAL DESCRIPTION

The ADM2482E/ADM2487E are isolated data transceivers with $\pm 15 \mathrm{kV}$ ESD protection and are suitable for high speed, halfduplex or full-duplex communication on multipoint transmission lines. For half-duplex operation, the transmitter outputs and receiver inputs share the same transmission line. Transmitter Output Pin Y is linked externally to Receiver Input Pin A, and Transmitter Output Pin Z to Receiver Input Pin B. The parts are designed for balanced transmission lines and comply with TIA/EIA- 485-A-98 and ISO 8482:1987(E).

The devices employ the Analog Devices, Inc., $i$ Coupler ${ }^{\circledR}$ technology to combine a 3-channel isolator, a three-state differential line driver, and a differential input receiver into a single package. An on-chip oscillator outputs a pair of square waveforms that drive an external transformer to provide isolated
power. The logic side of the device can be powered with either a 5 V or a 3.3 V supply, and the bus side is powered with an isolated 3.3 V supply.
The ADM2482E/ADM2487E driver has an active high enable, and the receiver has an active low enable. The driver output enters a high impedance state when the driver enable signal is low. The receiver output enters a high impedance state when the receiver enable signal is high.
The device has current-limiting and thermal shutdown features to protect against output short circuits and situations where bus contention might cause excessive power dissipation. The part is fully specified over the industrial temperature range of $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ and is available in a 16 -lead, wide-body SOIC package.

Rev. 0
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## ADM2482E/ADM2487E

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## REVISION HISTORY

5/08-Revision 0: Initial Version

## SPECIFICATIONS

All voltages are relative to their respective ground; $3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 1} \leq 5.5 \mathrm{~V}, 3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD} 2} \leq 3.6 \mathrm{~V}$. All minimum/maximum specifications apply over the entire recommended operation range, unless otherwise noted. All typical specifications are at $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD1}}=5 \mathrm{~V}$, $\mathrm{V}_{\mathrm{DD} 2}=3.3 \mathrm{~V}$, unless otherwise noted.

Table 1.


## ADM2482E/ADM2487E

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| TRANSFORMER DRIVER |  |  |  |  |  |  |
| Oscillator Frequency | fosc | 400 | 500 | 600 | kHz | $\mathrm{V}_{\mathrm{DD} 1}=5.0 \mathrm{~V}$ |
|  |  | 230 | 330 | 430 | kHz | $\mathrm{V}_{\mathrm{DD} 1}=3.3 \mathrm{~V}$ |
| Switch-On Resistance | RoN |  | 0.5 | 1.5 | $\Omega$ |  |
| Start-Up Voltage | VsTART |  | 2.2 | 2.5 | V |  |
| COMMON-MODE TRANSIENT IMMUNITY ${ }^{1}$ |  | 25 |  | $\mathrm{kV} / \mu \mathrm{S}$ | VCM $=1 \mathrm{kV}$, transient <br> magnitude $=800 \mathrm{~V}$ |  |

${ }^{1} \mathrm{CM}$ is the maximum common-mode voltage slew rate that can be sustained while maintaining specification-compliant operation. $\mathrm{V}_{\mathrm{CM}}$ is the common-mode potential difference between the logic and bus sides. The transient magnitude is the range over which the common-mode is slewed. The common-mode voltage slew rates apply to both rising and falling common-mode voltage edges.

## TIMING SPECIFICATIONS

$\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Table 2. ADM2482E

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DRIVER |  |  |  |  |  |  |
| Propagation Delay | $\mathrm{t}_{\text {DPLH, }}$ t ${ }_{\text {DPHL }}$ |  |  | 100 | ns | RDIFF $=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Output Skew | toskew |  |  | 8 | ns | RDIFF $=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Rise Time/Fall Time | $\mathrm{t}_{\mathrm{DR},} \mathrm{t}_{\mathrm{DF}}$ |  |  | 15 | ns | $\mathrm{R}_{\text {DIFF }}=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Enable Time | $\mathrm{tzL}^{\text {L }}$ tzH |  |  | 120 | ns | $R_{L}=110 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 5 and Figure 11 |
| Disable Time | tız, thz |  |  | 150 | ns | $\mathrm{R}_{\mathrm{L}}=110 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 5 and Figure 11 |
| RECEIVER |  |  |  |  |  |  |
| Propagation Delay | $\mathrm{t}_{\text {PLH, }} \mathrm{t}_{\text {PHL }}$ |  |  | 110 | ns | $C_{L}=15 \mathrm{pF}$, see Figure 6 and Figure 10 |
| Output Skew | $\mathrm{t}_{\text {skew }}$ |  |  | 8 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$, see Figure 6 and Figure 10 |
| Enable Time | tzL, tz ${ }_{\text {l }}$ |  |  | 13 | ns | $\mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega, \mathrm{C}_{L}=15 \mathrm{pF}$, see Figure 7 and Figure 12 |
| Disable Time | tız, thz |  |  | 13 | ns | $\mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$, see Figure 7 and Figure 12 |

Table 3. ADM2487E

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DRIVER |  |  |  |  |  |  |
| Propagation Delay | $\mathrm{t}_{\text {DPLH, }} \mathrm{t}_{\text {DPHL }}$ | 250 |  | 700 | ns | $\mathrm{R}_{\text {DIFF }}=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Output Skew | toskew |  |  | 100 | ns | $\mathrm{R}_{\text {DIFF }}=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Rise Time/Fall Time | $\mathrm{t}_{\mathrm{DR}}$, $\mathrm{t}_{\text {FF }}$ | 200 |  | 1100 | ns | R ${ }_{\text {DIFF }}=54 \Omega, C_{L}=100 \mathrm{pF}$, see Figure 4 and Figure 9 |
| Enable Time | $\mathrm{t}_{\mathrm{tL}}, \mathrm{t}_{\text {z }}$ |  |  | 2.5 | $\mu \mathrm{s}$ | $\mathrm{R}_{\mathrm{L}}=110 \Omega, \mathrm{C}_{L}=50 \mathrm{pF}$, see Figure 5 and Figure 11 |
| Disable Time | $\mathrm{t}_{\mathrm{Lz}}, \mathrm{t}_{\mathrm{Hz}}$ |  |  | 200 | ns | $\mathrm{R}_{\mathrm{L}}=110 \Omega, C_{L}=50 \mathrm{pF}$, see Figure 5 and Figure 11 |
| RECEIVER |  |  |  |  |  |  |
| Propagation Delay | tPLH, tPHL |  |  | 200 | ns | $C_{L}=15 \mathrm{pF}$, see Figure 6 and Figure 10 |
| Output Skew | tskew |  |  | 30 | ns | $\mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$, see Figure 6 and Figure 10 |
| Enable Time | $\mathrm{t}_{\mathrm{zL}}, \mathrm{t}_{\text {z }}$ |  |  | 13 | ns | $\mathrm{R}_{\mathrm{L}}=1 \mathrm{k} \Omega, \mathrm{C}_{\mathrm{L}}=15 \mathrm{pF}$, see Figure 7 and Figure 12 |
| Disable Time | $\mathrm{t}_{\mathrm{Lz}}, \mathrm{t}_{\text {Hz }}$ |  |  | 13 | ns | $R_{L}=1 \mathrm{k} \Omega, C_{L}=15 \mathrm{pF}$, see Figure 7 and Figure 12 |

## ADM2482E/ADM2487E

## PACKAGE CHARACTERISTICS

Table 4.

| Parameter | Symbol | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resistance (Input-Output) ${ }^{1}$ | R-O |  | $10^{12}$ |  | $\Omega$ |  |
| Capacitance (Input-Output) ${ }^{1}$ | Ci-O |  | 3 |  | pF | $\mathrm{f}=1 \mathrm{MHz}$ |
| Input Capacitance ${ }^{2}$ | $\mathrm{C}_{1}$ |  | 4 |  | pF |  |
| Input IC Junction-to-Case Thermal Resistance | $\theta_{\mathrm{JcI}}$ |  | 33 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | Thermocouple located at center of package underside |
| Output IC Junction-to-Case Thermal Resistance | $\theta$ лсо |  | 28 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ | Thermocouple located at center of package underside |

${ }^{1}$ This device is considered a 2-terminal device: Pin 1 to Pin 8 are shorted together and Pin 9 to Pin 16 are shorted together.
${ }^{2}$ Input capacitance is from any input data pin to ground.

## INSULATION AND SAFETY-RELATED SPECIFICATIONS

Table 5.

| Parameter | Symbol | Value | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: |
| Rated Dielectric Insulation Voltage |  | 2500 | V rms | 1-minute duration |
| Maximum Working Insulation Voltage |  | 560 | $\checkmark$ peak |  |
| Minimum External Air Gap (Clearance) | L(101) | 5.7 min | mm | Measured from input terminals to output terminals, shortest distance through air |
| Minimum External Tracking (Creepage) | L(102) | 6.1 min | mm | Measured from input terminals to output terminals, shortest distance along body |
| Minimum Internal Gap (Internal Clearance) |  | 0.017 min | mm | Distance through insulation |
| Tracking Resistance (Comparative Tracking Index) | CTI | >175 | V | DIN IEC 112/VDE 0303-1 |

## ADM2482E/ADM2487E

## TEST CIRCUITS



Figure 2. Driver Voltage Measurement


Figure 3. Driver Voltage Measurement


Figure 4. Driver Propagation Delay


Figure 5. Driver Enable/Disable


Figure 7. Receiver Enable/Disable


Figure 8. Supply-Current Measurement Test Circuit

## SWITCHING CHARACTERISTICS



Figure 9. Driver Propagation Delay, Rise/Fall Timing


Figure 10. Receiver Propagation Delay

Figure 11. Driver Enable/Disable Timing


Figure 12. Receiver Enable/Disable Timing

## ADM2482E/ADM2487E

## ABSOLUTE MAXIMUM RATINGS

All voltages are relative to their respective ground; $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted.

Table 6.

| Parameter | Rating |
| :---: | :---: |
| VDD1 | -0.5 V to +6 V |
| $V_{\text {DD2 }}$ | -0.5 V to +6 V |
| Digital Input Voltages (DE, $\overline{\mathrm{RE}}$, TxD) | -0.5 V to $\mathrm{V}_{\mathrm{DD} 1}+0.5 \mathrm{~V}$ |
| Digital Output Voltages |  |
| RxD | -0.5 V to $\mathrm{V}_{\mathrm{DD} 1}+0.5 \mathrm{~V}$ |
| D1, D2 | 13 V |
| Driver Output/Receiver Input Voltage | -9 V to +14 V |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Average Output Current per Pin | -35 mA to +35 mA |
| ESD (Human Body Model) on A, B, Y and $Z$ pins | $\pm 15 \mathrm{kV}$ |
| Lead Temperature |  |
| Soldering (10 sec) | $300^{\circ} \mathrm{C}$ |
| Vapor Phase (60 sec) | $215^{\circ} \mathrm{C}$ |
| Infrared (15 sec) | $220^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ESD CAUTION

|  | ESD (electrostatic discharge) sensitive device. <br> Charged devices and circuit boards can discharge <br> without detection. Although this product features <br> patented or proprietary protection circuitry, damage <br> may occur on devices subjected to high energy ESD. <br> Therefore, proper ESD precautions should be taken to <br> avoid performance degradation or loss of functionality. |
| :--- | :--- |

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Table 7. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
| :---: | :---: | :---: |
| 1 | D1 | Transformer Driver Terminal 1. |
| 2 | D2 | Transformer Driver Terminal 2. |
| 3 | $\mathrm{GND}_{1}$ | Ground, Logic Side. |
| 4 | $\mathrm{V}_{\mathrm{DD} 1}$ | Power Supply, Logic Side ( 3.3 V or 5 V ). Decoupling capacitor to $\mathrm{GND}_{1}$ required; capacitor value should be between $0.01 \mu \mathrm{~F}$ and $0.1 \mu \mathrm{~F}$. |
| 5 | RxD | Receiver Output Data. This output is high when $(A-B)>+200 \mathrm{mV}$ and low when $(A-B)<-200 \mathrm{mV}$. The output is tristated when the receiver is disabled, that is, when $\overline{\mathrm{RE}}$ is driven high. |
| 6 | $\overline{\mathrm{RE}}$ | Receiver Enable Input. This is an active-low input. Driving this input low enables the receiver; driving it high disables the receiver. |
| 7 | DE | Driver Enable Input. Driving this input high enables the driver; driving it low disables the driver. |
| 8 | TxD | Transmit Data. |
| 9 | $\mathrm{GND}_{2}$ | Ground, Bus Side. |
| 10 | NC | No Connect. This pin must be left floating. |
| 11 | Y | Driver Noninverting Output. |
| 12 | Z | Driver Inverting Output. |
| 13 | B | Receiver Inverting Input. |
| 14 | A | Receiver Noninverting Input. |
| 15 | $\mathrm{GND}_{2}$ | Ground, Bus Side. |
| 16 | $\mathrm{V}_{\mathrm{DD} 2}$ | Power Supply, Bus Side (Isolated 3.3 V Supply). Decoupling capacitor to $\mathrm{GND}_{2}$ required; capacitor value should be between $0.01 \mu \mathrm{~F}$ and $0.1 \mu \mathrm{~F}$. |

## ADM2482E/ADM2487E

## TYPICAL PERFORMANCE CHARACTERISTICS



Figure 14. ADM2487E IDDI Supply Current vs. Temperature (Data Rate $=500 \mathrm{kbps}, V_{D D 1}=5 \mathrm{~V}, V_{D D 2}=3.3 \mathrm{~V}, D E=1 \mathrm{~V}, \overline{R E}=0 \mathrm{~V}$ )


Figure 15. ADM2487E IDD2 Supply Current vs. Temperature (See Figure 8) (Data Rate $=500 \mathrm{kbps}, V_{D D 1}=5 \mathrm{~V}, V_{D D 2}=3.3 \mathrm{~V}, D E=1 \mathrm{~V}, \overline{R E}=0 \mathrm{~V}$ )


Figure 16. ADM2482E IDD1 Supply Current vs. Temperature (Data Rate $=$ $\left.16 \mathrm{Mbps}, V_{D D 1}=5 \mathrm{~V}, V_{D D 2}=3.3 \mathrm{~V}, D E=1 \mathrm{~V}, \overline{R E}=0 \mathrm{~V}\right)$


Figure 17. ADM2482E Supply Current vs. Temperature (See Figure 8) (Data Rate $\left.=16 \mathrm{Mbps}, V_{D D 1}=5 \mathrm{~V}, V_{D D 2}=3.3 \mathrm{~V}, D E=1, \overline{R E}=0 \mathrm{~V}\right)$


Figure 18. ADM2487E Driver Propagation Delay vs. Temperature


Figure 19. ADM2482E Driver Propagation Delay vs. Temperature


Figure 20. Output Current vs. Receiver Output High Voltage


Figure 21. Output Current vs. Receiver Output Low Voltage


Figure 22. Receiver Output High Voltage vs. Temperature $\left(I_{D D 2}=-4 m A\right)$


Figure 23. Receiver Output Low Voltage vs. Temperature $\left(l_{D D 2}=4 \mathrm{~mA}\right)$

$\mathrm{CH} 12.0 \mathrm{~V} \Omega \mathrm{CH} 22.0 \mathrm{~V} \Omega \underset{\substack{\mathrm{M} \\ 8.0 \mathrm{~ns} / \mathrm{pt}}}{\mathrm{M} 400 \mathrm{~ns} 125 \mathrm{MS} / \mathrm{s}} \mathrm{A} \mathrm{CH} 2 \leftrightharpoons 1.52 \mathrm{~V}$ 商
Figure 24. Switching Waveforms (50 $\Omega$ Pull-Up to $V_{D D 1}$ on D1 and D2)


Figure 25. Switching Waveforms (Break-Before-Make, $50 \Omega$ Pull-Up to $V_{D D 1}$ on D1 and D2)

## ADM2482E/ADM2487E



Figure 26. ADM2487E Driver/Receiver Propagation Delay, Low to High $\left(R_{\text {DIFF }}=54 \Omega, C_{L 1}=C_{L 2}=100 \mathrm{pF}\right)$


Figure 27. ADM2487E Driver/Receiver Propagation Delay, High to Low $\left(R_{\text {DIFF }}=54 \Omega, C_{L 1}=C_{L 2}=100 \mathrm{pF}\right.$ )


Figure 28. ADM2482E Driver/Receiver Propagation Delay, High to Low $\left(R_{\text {DIFF }}=54 \Omega, C_{L 1}=C_{L 2}=100 \mathrm{pF}\right)$


Figure 29. ADM2482E Driver/Receiver Propagation Delay, Low to High ( RIIFF $=54 \Omega, C_{L 1}=C_{L 2}=100 \mathrm{pF}$ )

## ADM2482E/ADM2487E

## CIRCUIT DESCRIPTION

## ELECTRICAL ISOLATION

In the ADM2482E/ADM2487E, electrical isolation is implemented on the logic side of the interface. Therefore, the part has two main sections: a digital isolation section and a transceiver section (see Figure 30). Driver input and data enable applied to the TxD and DE pins, respectively, and referenced to logic ground $\left(\mathrm{GND}_{1}\right)$ are coupled across an isolation barrier to appear at the transceiver section referenced to isolated ground $\left(\mathrm{GND}_{2}\right)$. Similarly, the receiver output, referenced to isolated ground in the transceiver section, is coupled across the isolation barrier to appear at the RxD pin referenced to logic ground.

## iCoupler Technology

The digital signals transmit across the isolation barrier using $i$ Coupler technology. This technique uses chip scale transformer windings to couple the digital signals magnetically from one side of the barrier to the other. Digital inputs are encoded into waveforms that are capable of exciting the primary transformer winding. At the secondary winding, the induced waveforms are decoded into the binary value that was originally transmitted.
Positive and negative logic transitions at the input cause narrow pulses ( $\sim 1 \mathrm{~ns}$ ) to be sent to the decoder, via the transformer. The decoder is bistable and is, therefore, either set or reset by the pulses, indicating input logic transitions. In the absence of logic transitions at the input for more than $\sim 1 \mu \mathrm{~s}$, a periodic set of refresh pulses indicative of the correct input state are sent to ensure dc correctness at the output. If the decoder receives no internal pulses for more than about $5 \mu \mathrm{~s}$, then the input side is assumed to be unpowered or nonfunctional, in which case the output is forced to a default state (see Table 8).


Figure 30. ADM2482E/ADM2487E Digital Isolation and Transceiver Sections

## TRUTH TABLES

The truth tables in this section use the abbreviations found in Table 8.

Table 8. Truth Table Abbreviations

| Letter | Description |
| :--- | :--- |
| H | High level |
| I | Indeterminate |
| L | Low level |
| X | Irrelevant |
| Z | High impedance (off) |
| NC | Disconnected |

Table 9. Transmitting

| Supply Status |  | Inputs |  | Outputs |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathbf{V}_{\text {DD } 1}$ | V $_{\text {DD } 2}$ | DE | TxD | Y | Z |
| On | On | H | H | H | L |
| On | On | H | L | L | H |
| On | On | L | X | Z | Z |
| On | Off | X | X | Z | Z |
| Off | On | L | X | Z | Z |
| Off | Off | X | X | Z | Z |

Table 10. Receiving

| Supply Status |  | Inputs |  | Outputs |
| :---: | :---: | :---: | :---: | :---: |
| $V_{\text {DD1 }}$ | VDD2 | A-B | $\overline{\mathbf{R E}}$ | RxD |
| On | On | $>-0.03 \mathrm{~V}$ | L or NC | H |
| On | On | <-0.2 V | L or NC | L |
| On | On | $-0.2 \mathrm{~V}<\mathrm{A}-\mathrm{B}<-0.03 \mathrm{~V}$ | L or NC | I |
| On | On | Inputs open | L or NC | H |
| On | On | X | H | Z |
| On | Off | X | L or NC | H |
| Off | Off | X | L or NC | L |

## ADM2482E/ADM2487E

## THERMAL SHUTDOWN

The ADM2482E/ADM2487E contain thermal shutdown circuitry that protects the part from excessive power dissipation during fault conditions. Shorting the driver outputs to a low impedance source can result in high driver currents. The thermal sensing circuitry detects the increase in die temperature under this condition and disables the driver outputs. This circuitry is designed to disable the driver outputs when a die temperature of $150^{\circ} \mathrm{C}$ is reached. As the device cools, the drivers are re-enabled at a temperature of $140^{\circ} \mathrm{C}$.

## TRUE FAIL-SAFE RECEIVER INPUTS

The receiver inputs have a true fail-safe feature that ensures that the receiver output is high when the inputs are open or shorted. During line-idle conditions, when no driver on the bus is enabled, the voltage across a terminating resistance at the receiver input decays to 0 V . With traditional transceivers, receiver input thresholds specified between -200 mV and +200 mV mean that external bias resistors are required on the $A$ and $B$ pins to ensure that the receiver outputs are in a known state. The true fail-safe receiver input feature eliminates the need for bias resistors by specifying the receiver input threshold between -30 mV and -200 mV . The guaranteed negative threshold means that when the voltage between A and B decays to 0 V , the receiver output is guaranteed to be high.

## MAGNETIC FIELD IMMUNITY

The limitation on the magnetic field immunity of the $i$ Coupler is set by the condition in which an induced voltage in the receiving coil of the transformer is large enough to either falsely set or reset the decoder. The following analysis defines the conditions under which this may occur. The 3 V operating condition of the ADM2482E/ADM2487E is examined because it represents the most susceptible mode of operation.
The pulses at the transformer output have an amplitude greater than 1 V . The decoder has a sensing threshold of about 0.5 V , thus establishing a 0.5 V margin in which induced voltages can be tolerated.

The voltage induced across the receiving coil is given by

$$
V=\left(\frac{-d \beta}{d t}\right) \sum \pi r_{n}^{2} ; n=1,2, \ldots, N
$$

where:
$\beta$ is the magnetic flux density (gauss).
$N$ is the number of turns in the receiving coil. $r_{n}$ is the radius of the $\mathrm{n}^{\text {th }}$ turn in the receiving coil (cm).

Given the geometry of the receiving coil and an imposed requirement that the induced voltage is, at most, $50 \%$ of the 0.5 V margin at the decoder, a maximum allowable magnetic field can be determined using Figure 31.


Figure 31. Maximum Allowable External Magnetic Flux Density
For example, at a magnetic field frequency of 1 MHz , the maximum allowable magnetic field of 0.2 kgauss induces a voltage of 0.25 V at the receiving coil. This is about $50 \%$ of the sensing threshold and does not cause a faulty output transition. Similarly, if such an event occurs during a transmitted pulse and is the worst-case polarity, it reduces the received pulse from $>1.0 \mathrm{~V}$ to 0.75 V , still well above the 0.5 V sensing threshold of the decoder.

Figure 32 shows the magnetic flux density values in terms of more familiar quantities, such as maximum allowable current flow at given distances away from the ADM2482E/ADM2487E transformers.


Figure 32. Maximum Allowable Current for Various Current-to-ADM2482E/ADM2487E Spacings

With combinations of strong magnetic field and high frequency, any loops formed by PCB traces could induce error voltages large enough to trigger the thresholds of succeeding circuitry. Care should be taken in the layout of such traces to avoid this possibility.

## APPLICATIONS INFORMATION

## PRINTED CIRCUIT BOARD LAYOUT

The isolated RS-485 transceiver of the ADM2482E/ADM2487E requires no external interface circuitry for the logic interfaces. Power supply bypassing is required at the input and output supply pins (see Figure 33).
Bypass capacitors are most conveniently connected between Pin 3 and Pin 4 for $V_{D D 1}$ and between Pin 15 and Pin 16 for $\mathrm{V}_{\mathrm{DD} 2}$. The capacitor value must be between $0.01 \mu \mathrm{~F}$ and $0.1 \mu \mathrm{~F}$. The total lead length between both ends of the capacitor and the input power supply pin must not exceed 20 mm .
Bypassing Pin 9 and Pin 16 is also recommended unless the ground pair on each package side is connected close to the package.


Figure 33. Recommended Printed Circuit Board Layout
In applications involving high common-mode transients, care must be taken to ensure that board coupling across the isolation barrier is minimized. Furthermore, the board layout must be designed such that any coupling that does occur equally affects all pins on a given component side.
Failure to ensure this can cause voltage differentials between pins exceeding the absolute maximum ratings of the device, thereby leading to latch-up or permanent damage.

## ISOLATED POWER SUPPLY CIRCUIT

The ADM2482E/ADM2487E integrate a transformer driver that, when used with an external transformer and linear voltage regulator (LDO), generates an isolated 3.3 V power supply to be supplied between $V_{\text {DD } 2}$ and $\mathrm{GND}_{2}$, as shown in Figure 34.

Pin D1 and Pin D2 of the ADM2482E/ADM2487E drive a center-tapped Transformer T1. A pair of Schottky diodes and a smoothing capacitor are used to create a rectified signal from the secondary winding. The ADP1710 LDO provides a regulated 3.3 V power supply to the ADM2482E/ ADM2487E bus-side circuitry ( $\mathrm{V}_{\mathrm{DD} 2}$ ).
When the ADM2482E/ADM2487E are powered by 3.3 V on the logic side, a step-up transformer is required to compensate for the forward voltage drop of the Schottky diodes and the voltage drop across the regulator. The transformer turns ratio should be chosen to ensure just enough headroom for the ADP1710 LDO to output a regulated 3.3 V output under all operating conditions.
If the ADM2482E/ADM2487E are powered by 5 V on the logic side, then a step-down transformer should be used. For optimum efficiency, the transformer turns ratio should be chosen to ensure just enough headroom for the ADP1710 LDO to output a regulated 3.3 V output under all operating conditions.


## ADM2482E/ADM2487E

## TYPICAL APPLICATIONS

Figure 35 and Figure 36 show typical applications of the ADM2482E/ADM2487E in half-duplex and full-duplex RS-485 network configurations. Up to 256 transceivers can be connected to the RS- 485 bus. To minimize reflections, the
line must be terminated at the receiving end in its characteristic impedance, and stub lengths off the main line must be kept as short as possible. For half-duplex operation, this means that both ends of the line must be terminated, because either end can be the receiving end.


Figure 35. ADM2482E/ADM2487E Typical Half-Duplex RS-485 Network


Figure 36. ADM2482E/ADM2487E Typical Full-Duplex RS-485 Network

## OUTLINE DIMENSIONS



Figure 37. 16-Lead Standard Small Outline Package [SOIC_W] Wide Body
(RW-16)
Dimensions shown in millimeters and (inches)
ORDERING GUIDE

| Model | Data Rate (Mbps) | Temperature Range | Package Description | Package Option |
| :--- | :--- | :--- | :--- | :--- |
| ADM2482EBRWZ ${ }^{1}$ | 16 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead SOIC_W | RW-16 |
| ADM2482EBRWZ-REEL71 | 16 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead SOIC_W | RW-16 |
| ADM2487EBRWZ $^{1}$ | 0.5 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead SOIC_W | RW-16 |
| ADM2487EBRWZ-REEL7 $^{1}$ | 0.5 | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 16 -Lead SOIC_W | RW-16 |

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## ADM2482E/ADM2487E

NOTES

NOTES

## ADM2482E/ADM2487E

## NOTES


[^0]:    ${ }^{1} Z=$ RoHS Compliant Part.

